PMDC Motor Parameter Estimation using Bio-Inspired Optimization Algorithms

V. Sankardoss, Member, IEEE, and P. Geethanjali, Member, IEEE

Abstract-The precise estimation of the motor parameter is essential to design the appropriate controller. The main goal of this paper is to estimate the parameters of permanent magnet DC (PMDC) motor used in a wheelchair, applying standard as well as a dynamic particle swarm optimization (PSO), ant colony optimization (ACO) and artificial bee colony (ABC) along with experimental methods. The electromechanical, mechanical and electrical parameters such as torque constant, back-emf constant, moment of inertia, viscous friction coefficient, armature inductance and resistance are estimated using both the experimental and optimization methods. The motor is modeled in Matlab/Simulink R2015a using the estimated motor parameters and studied the performance with different loading condition starting from no-load to full-load. The simulated results of motor performance with estimated parameters are compared with the experimental load test results. The results showed that the PMDC motor parameters estimated from dynamic PSO with varying inertia weight as well as artificial bee colony algorithm have comparatively very less speed and current error than standard PSO, dynamic PSO with constant inertia weight, and ant colony optimization algorithms. Further, parameters from dynamic PSO with varying inertia weight showed speed as well as current error less than 0.5% and the artificial bee colony algorithm shown current error slightly more than 0.5%. However, analysis of variance (ANOVA) tests shown no significant difference in current and speed performance with parameter estimated from artificial bee colony and dynamic PSO with varying inertia weight. Further, artificial bee colony algorithm convergence is faster than dynamic PSO with varying inertia weight. But parameters estimated from dynamic PSO with varying inertia weight are precise and may be appropriate for the design of the motor controllers.

Index Terms— Ant colony optimization, artificial bee colony, particle swarm optimization, PMDC motor, wheelchair

I. INTRODUCTION

S everal types of DC motors have been widely used in home and industrial applications. Advances in permanent magnet materials and desirable features such as light weight, low cost, low speed, etc. augmented the applications of permanent magnet DC (PMDC) motor particularly in wheelchair drive [1]-[2]. The accurate design of wheelchair

"This work was supported by the Science and Engineering Research Board (SERB), Government of India, SB/FTP/ETA-54/2013".

V. Sankardoss, School of Electrical Engineering, VIT University, Vellore, India (e-mail: vsankardoss@vit.ac.in).

Dr. P. Geethanjali, School of Electrical Engineering, VIT University, Vellore, India (e-mail: pganjali78@hotmail.com).

drive controller is essential to provide safety and comfort to the wheelchair user [3]-[6]. However, the design of the precise controller for the motor using the conventional proportionalintegral-derivative (PID) technique, intelligent control technique such as neural network, fuzzy, neuro-fuzzy, etc. require accurate modeling of the motor that considers the nonlinear dynamics [7]-[9]. The estimation of motor parameters is also essential for condition monitoring, fault diagnosis, etc. However, the identification of non-linear dynamics is very complex and design of controller requires accurate parameters to model the motor. The poor estimation of motor parameters may lead to the controller with suboptimal performance and may eventually lead to instability and deterioration of the system.

1

Researchers have attempted different methods of parameter identification for various types of DC motors. Different approaches are widely developed for parameter estimation of DC motor [10]-[19]. However, in the case of PMDC motor not much parameter estimation techniques are reported in the literature. Techniques like frequency response method [20], gradient method [21], quantized identification method [22], recursive least squares method [23] are employed to estimate PMDC motor parameters.

The experimental method of parameter estimation is a difficult problem and requires the knowledge of the relationship between the parameter and environmental factors. The cost and time involved in experimental parameter estimation are high for setting up an experiment with necessary sensors, data acquisition equipment etc. Recently, the optimization algorithm gained interest in the system parameter identification due to the computational power of the personal computer. Researchers have developed genetic algorithm [24], adaptive tabu search technique [25], particle swarm optimization [26], etc. for parameter estimation of different types of motors. The bio-inspired optimization techniques have the advantage of finding global optimal by some operations to fit the objective function. Therefore, bio-inspired optimization techniques are preferred.

The authors have attempted particle swarm optimization (PSO) algorithm using the standard as well as dynamic approaches, ant colony optimization (ACO) and artificial bee colony (ABC) for estimation of PMDC motor parameters used in a wheelchair. The results of parameters estimated from

optimization methods are validated with the experimental load test data. Furthermore, the performance of parameters estimated from the bio-inspired optimization techniques and experimental method are compared.

In this paper, section 2 describes the modeling of PMDC motor using electromechanical energy conversion principle. Section 3 details the experimental method of parameter estimation as well as bio-inspired optimization techniques. The results of various methods are discussed in section 4. Finally, the conclusion has been presented in section 5.

II. MATHEMATICAL MODEL OF PMDC MOTOR

The dynamic equations of a PMDC motor can be modeled based on the Kirchhoff's voltage law and the Newton's moment law using equations (1) - (4).

$$v_{arm} = L_{arm} \frac{di_{arm}}{dt} + R_{arm} i_{arm} + e_b \tag{1}$$

$$T_e = J \frac{d\omega}{dt} + f\omega + T_l \tag{2}$$

$$T_e = K_t i_{arm} \tag{3}$$

$$e_b = \omega K_b \tag{4}$$

The PMDC motor inputs are DC input voltage (v_{arm}) and load torque (T_l) . The outputs are the angular speed (ω) and the armature current (i_{arm}) . The block diagram representation of PMDC motor model from Laplace transform of equations (1) - (4) is shown in Fig. 1.





III. MOTOR PARAMETER ESTIMATION

In order to model the motor, the parameters are estimated using experimental techniques as well as using bio-inspired optimization techniques. The subsequent section discusses the motor parameter estimation.

A. Experimental Techniques

The parameters of motor include electrical parameters armature resistance, armature inductance; mechanical parameters - viscous friction coefficient, moment of inertia; and electromechanical constants - back-emf constant and torque constant which define the electrical to mechanical energy conversion. The parameters are estimated from experimental techniques without considering losses and temperature effect due to computation complexity.

1) Determination of armature resistance (R_{arm})

The determination of armature resistance in PMDC motor is complex. The usage of multi-meter to measure the armature resistance is typically very high owing to the connection between the brush and commutator along with the internal resistance of multi-meter. Moreover, the digital milliohmmeter also fails to measure the precise armature resistance. The stall test process has been proven to be a reliable technique and hence is employed for measurement of resistance.

2

A DC voltage is supplied to the motor, which restricts the armature rotation and draws 10% of the rated current or slightly more. Besides, the temperature of the winding also influences the resistance of the armature and not considered due to complexity. The voltage across the armature and armature current are measured for five different voltages when subjected to no-load conditions swiftly.

The electrical equation (1) is reduced to

$$v_{arm} = i_{arm} R_{arm} \tag{5}$$

At steady state, the voltage across the inductance is zero and brush drop is negligible. The armature resistance is calculated from the linear slope of the armature voltage and armature current.

2) Determination of armature inductance (L_{arm})

The armature inductance may be measured using two techniques; LCR (Inductance, Capacitance, and Resistance) meter and impedance method. However, the impedance method permits the measurement with changing field current. The field current is not possible to vary in PMDC to limit the armature current to one-tenth of the rated current to minimize the armature reaction. Therefore, in this work, the inductance is measured using an LCR meter.

3) Determination of back-emf constant (K_b)

The motor is supplied with a DC voltage enabling the armature rotation. The armature current, armature voltage, and speed are measured for different supply voltage across the armature under a no-load condition.

The back-emf (e_b) is calculated for different supply voltage using equation (1). Further, the back-emf of PMDC motor is proportional to an angular speed of the shaft and is calculated using equation (4) at steady state. The back-emf constant is estimated from the slope of the characteristics between backemf and angular speed.

4) Determination of torque constant (K_t)

Under steady state condition, the electrical power (P_e) is equal to mechanical power (P_m) .

$$P_{\rm e} = P_{\rm m} \tag{6}$$

The electrical power is calculated from the experimental measurements using the equation (7).

$$P_{\rm e} = (v_{\rm arm} - i_{\rm arm} R_{\rm arm}) i_{\rm arm} \tag{7}$$

The electromagnetic torque (T_e) is determined from the mechanical power and angular speed from the equation (8).

$$T_e = \frac{P_m}{\omega} \tag{8}$$

The electrical power and hence the electromagnetic torque are calculated from the equation (7) and (8). The torque constant is determined from the slope of the characteristics between electromagnetic torque and armature current.

5) Determination of viscous friction coefficient (f)

Under the steady-state condition, the viscous friction coefficient (f) is determined from the linear slope of the characteristics between electromagnetic torque and angular speed using equation (9).

$$T_e = f\omega \tag{9}$$

6) Determination of moment of inertia (J)

To determine the moment of inertia of PMDC motor, retardation test is conducted. The PMDC machine is made to run at a speed just above the rated speed of the motor. Then the supply to the armature is cut off and the motor is allowed to reach the zero speed. The torque losses are supplied with energy stored in the moment of inertia. Therefore, the moment of inertia can be calculated using equation (10).

$$J = \frac{f\omega + A}{d\omega/dt} \tag{10}$$

Time for speed fall is recorded and the graph is plotted between the angular speed and time to obtain the slope of retardation curve.

B. Optimization Techniques

The experimental method of motor parameter estimation is time-consuming, complex and expensive for setting up the experiment. Nevertheless, precise estimation is essential for motor parameters during the design of the controller for the wheelchair. Figure 2 shows the block diagram of PMDC motor parameter estimation using optimization algorithms.



Fig. 2. Estimation of PMDC motor parameters using optimization algorithm

The PMDC motor parameter estimation is based on normalized angular speed error and current error, which compares the angular speed and current response of the real system with the response of PMDC motor model using estimated parameters.

The response of the real system is given by equation (11).

$$y = h(z, u)$$
 (11)

where

where

The response of estimated parameter model is given by equation (12).

ŷ

$$= h\left(\hat{z}, u\right) \tag{12}$$

$$\hat{\vec{z}} = \begin{bmatrix} \vec{R}_{arm} & \hat{L}_{arm} & \hat{K}_{b} & \hat{K}_{t} & \hat{f} & \hat{J} \end{bmatrix}$$

$$\hat{\vec{y}} = \begin{bmatrix} \hat{\omega} & \hat{i}_{arm} \end{bmatrix}$$

The fitness evaluation function is based on mean square error calculated from normalized angular speed and normalized armature current response *y* of the real system and normalized estimated angular speed and normalized armature current response \hat{y} of PMDC motor model using equation (13).

$$F(z) = F_1(z) + aF_2(z)$$
(13)

$$F_1(\hat{z}) = \frac{1}{N} \sum_{j=1}^N \left\| \omega - \hat{\omega} \right\|^2$$
 and
$$F_2(\hat{z}) = \frac{1}{N} \sum_{j=1}^N \left\| i_{am} - i_{am}^{\wedge} \right\|^2$$

The objective function $F(\hat{z})$ is equal to zero, only when $\hat{\omega} = \omega$ and $\hat{i}_{arm} = \hat{i}_{arm}$ for *N* number of samples.

In this work, the authors attempted to minimize the objective function to estimate appropriate parameters of PMDC motor using the standard as well as dynamic particle swarm optimization algorithm, ant colony optimization algorithm, and artificial bee colony algorithm.

1) Standard Particle Swarm Optimization Algorithm

The particle swarm optimization (PSO) is one of the evolutionary population-based optimization techniques inspired by the behavior of bird flocks, fish schools etc. The PSO begins with the population of individuals called particles [27]-[31]. In PSO, each particle constitutes a number of parameters to be optimized known as a candidate solution in a multidimensional space. There are six parameters R_{arm} , L_{arm} , K_b , K_b , f and J to be estimated and therefore each particle is a point in six-dimensional search space. The population of particles is called the swarm. The PSO starts with the random initialization of swarm size and particle of the swarm. The swarm searches the optimal solution in a multidimensional parameter space starting with random position and zero velocity.

The swarm moves in the search space depending on the fitness value estimated from the defined objective function. The particles in the swarm drive toward the best solution by adjusting the velocity based on own experience and other particle experience known as neighborhood solution at every time step in the parameter search solution. The velocity and hence the position of each particle are updated over a time in a number of iterations by evaluating the fitness function and comparing current solution with *pbest* and *gbest*. Mathematically the velocity and position of each particle *i*, in *j* dimensional parameter space are calculated using the equation (14)-(15).

$$v_{ij}^{k+1} = w^* v_{ij}^k + c_1^* rand_1^* (pbest_{ij}^k - x_{ij}^k) + c_2^* rand_2^* (gbest_j^k - x_{ij}^k)$$
(14)
$$x_{ij}^{k+1} = x_{ij}^k + v_{ij}^{k+1}$$
(15)

The equation (14) constitutes of a momentum of the particle (w), iterative component (c_1) , and the social component (c_2) . The momentum of the particle is a function of inertia parameter, provides the access for the particle to move in the search space with the velocity of the particle v_{ij} . The iterative component otherwise known as the cognitive component, is based on the distance between the own best position (*pbest*) with the current position i.e. own distance. The social component is a function reflects the social behavior of the particle by considering the distance between best positions in a swarm (*gbest*) with the current position. The iterative and social components change the velocity of the particle and avoid hitting the boundary.

The six parameters, R_{arm} , L_{arm} , K_b , K_b , f, and J to be estimated is a particle and treated as a point in a six dimensional space. For the PSO algorithm, the i^{th} particle is represented as { R_{armb} , L_{armi} , K_{bi} , K_{ti} , f_i , J_i }. The position vector in PSO is given by equation (16).

$$\begin{bmatrix} R_{ami} & L_{ami} & K_{bi} & K_{ii} & f_{i} & J_{i} \end{bmatrix} = \begin{bmatrix} R_{am1}^{1} & L_{ami1}^{1} & K_{bi}^{1} & K_{bi}^{1} & K_{i}^{1} & f_{i}^{1} & J_{i}^{1} \\ R_{am2}^{1} & L_{am2}^{1} & L_{am2}^{1} & K_{b2}^{1} & K_{i2}^{1} & f_{2}^{1} & J_{2}^{1} \\ \cdots & \cdots & \cdots & \cdots & \cdots \\ R_{amid}^{1} & L_{amid}^{1} & L_{amid}^{1} & K_{b(M-1)}^{1} & K_{i(M-1)}^{1} & f_{(M-1)}^{1} & J_{M}^{1} \\ R_{amid}^{1} & L_{amid}^{1} & K_{bM}^{1} & K_{bM}^{1} & K_{iM}^{1} & f_{M}^{1} & J_{M}^{1} \end{bmatrix}$$

$$(16)$$

The response angular speed (ω) and armature current (i_a) are estimated using the armature voltage, load torque and current iterative parameter $\{R_{armi}^k, L_{armi}^k, K_{bi}^k, K_{ii}^k, f_i^k, J_i^k\}$ in a PMDC motor model. The fitness function (13) is evaluated to find the best position for the *i*th particle (*pbest*) and the best position of the swarm (*gbest*) using normalized angular speed and armature current.

In this work, the positions of the i^{th} particle are random sequences, which are limited in the ranges $[x_{min}, x_{max}]$ using equation (17).

$$\begin{cases} R_{armi \ (\min)} \leq R_{armi} \leq R_{armi \ (\max)} \\ L_{armi \ (\min)} \leq L_{armi} \leq L_{armi \ (\max)} \\ K_{bi \ (\min)} \leq K_{bi} \leq K_{bi \ (\max)} \\ K_{ii \ (\min)} \leq K_{ii} \leq K_{ii \ (\max)} \\ f_{i \ (\min)} \leq f_{i} \leq f_{i \ (\max)} \\ J_{i \ (\min)} \leq J_{i} \leq J_{i \ (\max)} \end{cases}$$
(17)

These conditions strongly depend on the user's experience as well as the problem considered.

The best position found for the i^{th} particle is represented as { $pbestR_{armi}^{k}, pbestL_{armi}^{k}, pbestK_{bi}^{k}, pbestK_{ti}^{k}, pbestf_{i}^{k}, pbestJ_{i}^{k}$ }. The best position found by the swarm is represented as $\{gbestR_{arm}^{k}, gbestL_{arm}^{k}, gbestK_{b}^{k}, gbestK_{t}^{k}, gbestf^{k}, gbestJ^{k}\}.$ The parameter vector $\{R_{armi}^k, L_{armi}^k, K_{bi}^k, K_{ti}^k, f_i^k, J_i^k\}$ is set to { $pbestR_{armi}^k$, $pbestL_{armi}^k$, $pbestK_{bi}^k$, $pbestK_{ii}^k$, $pbestf_i^k$, $pbestJ_i^k$ } if the current iteration (k) parameter vector is better than previous iteration (k-1) pbest parameter vector. Similar to *pbest*, the parameter vector $\{R_{armi}^k, L_{armi}^k, K_{bi}^k, K_{ti}^k, f_i^k, J_i^k\}$ is set to { $gbestR_{arm}^{k}, gbestL_{arm}^{k}, gbestK_{h}^{k}, gbestK_{t}^{k}, gbestf^{k}, gbestJ^{k}$ } if the current iteration (k) parameter vector is better than previous iteration (k-1) gbest parameter vector. The velocity of the *i*th particle is represented as { v_{Rarmi} , v_{Larmi} , v_{Kbi} , v_{Ki} , v_{fi} , v_{Ji} }. The velocity vector is calculated using (14) and the position of the i^{th} particle is updated through (15). The speed of convergence and number of iterations depends on the initialization range. This process is repeated until the userdefined goal is met. In this parameter estimation problem, the goal is that the k^{th} current iteration number reaches the maximum iteration number.

The solution of the parameter estimated at the end of iteration.

$$\begin{bmatrix} \hat{R}_{arm} & \hat{L}_{arm} & \hat{K}_{b} & \hat{K}_{t} & \hat{f} & \hat{J} \end{bmatrix} = \begin{bmatrix} gbest_{R_{arm}^{*}} & gbest_{L_{arm}^{*}} & gbest_{K_{b}^{*}} & gbest_{K_{t}^{*}} & gbest_{f^{*}} & gbest_{f^{*}} \end{bmatrix}$$
(18)

The standard PSO algorithm is obviously one of the simplest optimization algorithms. However, parameter convergence may not occur in standard PSO and affect the design of the controller. Therefore, the motor parameters are also estimated using dynamic PSO. In standard PSO, motor parameters are estimated for $c_1 = c_2 = 1$ and w = 1.

2) Dynamic Particle Swarm Optimization Algorithm

A dynamic particle swarm optimization (PSO) algorithm is based on time-varying cognitive (c_1) and the social component (c_2) with or without varying inertia weight. It is desirable to encourage searching the solution through the entire search space without trapping around local solution as well as making particle convergence towards the global solution. Proper control of c_1 and c_2 in addition to inertia weight w, will help to reach the optimal solution in an efficient way in PSO with varying inertia weight. It may be possible for individuals not get trapped in local minima at an early stage and converge towards the global solution at the latter stage using the iterative cognitive, social parameters with constant or varying inertia weight.

The time-varying representation of c_1 , c_2 and w are given by equations (19)-(21).

$$c_{1} = \left(c_{1Fv} - c_{1Iv}\right) \frac{k}{number \ of \ iteration} + c_{1Iv}$$
(19)

$$c_2 = (c_{2Fv} - c_{2Iv}) \frac{k}{number of iteration} + c_{2Iv}$$
(20)

This article has been accepted for publication in a future issue of this journal, but has not been fully edited. Content may change prior to final publication. Citation information: DOI 10.1109/ACCESS.2017.2679743, IEEE

$$w = (w_1 - w_2) \frac{(number \ of \ iteration \ -k)}{number \ of \ iteration} + w_2$$
(21)

The flowchart of the standard PSO, as well as dynamic PSO algorithms for estimation of PMDC motor parameters, is shown in Fig. 3.



Fig. 3. Flowchart of the standard as well as dynamic PSO algorithm

3) Artificial Bee Colony Algorithm

In 2005, Karaboga [32-36] has developed an artificial bee colony (ABC) algorithm, a population-based search technique from scrounging behavior of bees for solving optimization problems. In ABC algorithm, the bees are divided as employed bees, onlooker bees, and scout bees.

The employed bees find the food sources position and share the information to onlooker bees at the hive. On the other hand, onlooker bees select the high-quality food sources based on nectar information and search further around the selected food sources. Scout bees independently search the new food sources to replace the abandoned food sources of employed bees.

5

The ABC algorithm begins with the initial population of food source positions (S_p) . The *i*th food source is defined with the *d*-dimensional vector $P_i = [p_{i,1}, p_{i,2}, ..., p_{i,d}]$ for $i = 1, 2, ..., S_p$. Each food source position/solution is generated using equation (22).

$$p_{ij} = p_{\min, j} + rand \ (0,1) \left(p_{\max, j} - p_{\min, j} \right)$$
(22)
where $j = 1, 2, ..., d$.

In ABC algorithm, each food source position corresponds to six PMDC motor parameters to be estimated. The position of the food sources is limited using equation (17). After initialization, all employed bees search for the food sources and generate candidate food source position/solution using equation (23).

$$v_{ij} = p_{ij} + \phi_{ij} (p_{ij} - p_{gj})$$
(23)

where $g = 1, 2, ..., S_p$ and ϕ_{ij} is a random value in the range [-1, 1].

After generation of new food source position, the fitness of the new food source position is evaluated. The employed bees would replace the previous food sources position with new one; if the fitness value of the new food sources is better otherwise the employed bees retain the previous food source position. The employed bees share food source position and nectar information to onlooker bees.

An onlooker bees select the food sources depending on the probability value estimated using equation (24).

$$pr_{i} = \frac{fit_{i}}{\sum_{i=1}^{S_{p}} fit_{i}}$$
(24)

where fit_i is the fitness value of i^{th} food source position which depends on food source position. The number of food source position S_p is equal to the number of employed bees/onlooker bees.

The food source position is abandoned in case no improvement in the food source position is observed for predetermined number of cycles. Subsequently, scout bees discover the new food source position using equation (22). The new food source discovered by the scout bees will replace the abandoned one. This process of identification of best food source position is continued until the termination criteria is reached or the maximum number of iteration is reached. Figure 4 shows the flowchart of the ABC algorithm for estimation of PMDC motor parameters.



Fig. 4. Flowchart of the ABC algorithm

4) Ant Colony Optimization Algorithm

The Dorigo [37-40] has proposed ant colony optimization (ACO) from the inspiration of ant, to find an optimal path between food and nest. The optimal solution is found via the amount of pheromone on the ground. This concept has been used to estimate optimal value of PMDC motor parameter. The parameters are limited in the range and are given by equation (17). Each parameter is the vector corresponds to a layer/level. The upper and lower limits of the parameter depend on the user experience and divided into q number of nodes with possible values.

Ant colony optimization algorithm begins with initialization of ants and equal amounts of pheromone trail. The number of layers is six, constituting a motor parameter in each layer. Each layer consists of q nodes with permissible values assigned to each node using equation (17).

At each iteration, ant assumes the path using equation (25) to construct the probabilistic state transition rule for a complete solution. The state transition rule is mainly based on the state of pheromone.

$$p_{qs} = \frac{\tau_{qs}^{\alpha}}{\sum_{q=1}^{d} \tau_{qs}^{\alpha}}$$
(25)

The objective function is evaluated corresponding to the complete path to determine best and worst path of H ants. Subsequently, the optimal solution is obtained when all the ants follow the same best path.

If optimal solution is not obtained, the pheromone information is updated using equation (26).

$$\tau_{qs} = (1 - \rho)\tau_{qs} + \frac{Q}{f_{best}}$$
(26)

with the updated pheromones, the process continues until the end of the iteration is reached. Figure 5 presents the flowchart of the ACO algorithm for estimation of PMDC motor parameters.



IV. RESULTS AND DISCUSSION

In all algorithms of parameter estimation, the maximum number of iteration is set to 45, population size is set to 50, and parameters are calculated from the average of 30 runs. The other specific parameters of algorithms are:

1) PSO settings:

The PSO parameter considered for simulation is shown in Table 1. It can be observed that there is a difference among the standard PSO, dynamic PSO with constant inertia weight and dynamic PSO with varying inertia weight algorithms.

2169-3536 (c) 2016 IEEE. Translations and content mining are permitted for academic research only. Personal use is also permitted, but republication/redistribution requires IEEE permission. See http://www.ieee.org/publications_standards/publications/rights/index.html for more information.

6

This article has been accepted for publication in a future issue of this journal, but has not been fully edited. Content may change prior to final publication. Citation information: DOI 10.1109/ACCESS.2017.2679743, IEEE Access

The cognitive and social parameters are time-varying variables in the velocity update equation of the dynamic PSO algorithm with constant inertia weight. Large cognitive and small social parameters are used in the beginning to enhance the global search and then small cognitive and large social parameters are used at the end to improve the convergence of the algorithm. Further, dynamic PSO is accelerated with varying inertia along with varying c_1 and c_2 .



7

Dynamic PSO with varying inertia weight algorithm (c)

TABLE I PARTICLE SWARM OPTIMIZATION PARAMETERS

	initial particles	initial	inertia weight, w	acceleration	independent					
Algorithms	positions	particles		coefficients,	random sequences,					
		velocities		c_1 and c_2	r_1 and r_2					
Standard PSO	random numbers	0	w = constant = 1	$c_1 = c_2 = \text{constant} = 1$	random numbers					
	€ (0, 1)				€ (0, 1)					
Dynamic PSO	random numbers	0	w = constant = 1	$c_1 = \text{varying} = 2 \text{ to } 0.1$	random numbers					
with constant	€ (0, 1)			$c_2 = \text{varying} = 0.1 \text{ to } 2$	€ (0, 1)					
inertia weight										
Dynamic PSO	random numbers	0	w = varying =	$c_1 = \text{varying} = 2 \text{ to } 0.1$	random numbers					
with varying	€ (0, 1)		0.9 to 0.4	$c_2 = \text{varying} = 0.1 \text{ to } 2$	€ (0, 1)					
inertia weight										

2.08

2.04 20

2) ABC settings:

The ϕ_{ij} is a random value and chosen as 0.5.

3) ACO settings:

The value of pheromone α is set to 1, a number of nodes are 100, and range of evaporation rate (ρ) is between 0 and 1. In this work, ρ is chosen as 0.2.

The convergence of the standard PSO, dynamic PSO with constant inertia weight and dynamic PSO with varying inertia weight algorithms, ACO and ABC against the iteration step number is shown in Fig. 6. The average convergence value and a number of iterations for optimization algorithms for 30 runs are shown in Table 2.







Fig. 6. Best fitness value versus number of iterations

TABLE II

AVERAGE CONVERGENCE VALUE AND NUMBER OF **ITERATION FOR 30 RUNS**

	Value of final	Number of					
Algorithms	convergence	iterations					
Standard PSO	5.8102	25.68					
Dynamic PSO with	6.3889	30.46					
constant inertia weight							
Dynamic PSO with	5.1957	21.63					
varying inertia weight							
Artificial bee colony	7.4883	11.76					
Ant colony optimization	1.9381	6.16					

Table 3 shows the estimated PMDC motor parameters using standard PSO, dynamic PSO with constant inertia weight and dynamic PSO with varying inertia weight algorithms, ABC and ACO algorithms. To determine the appropriate estimated parameters, the motor is simulated with varying load conditions and compared the results with actual load test results.

AVERAGE OF ESTIMATED PMIDC MOTOR PARAMETERS FOR 50 KUNS										
Parameters	Experimental	Standard	Dynamic PSO	Dynamic PSO	Artificial	Ant colony				
	tests	PSO	with constant	with varying	bee colony	optimization				
			inertia weight	inertia weight						
R_a, Ω	0.162	0.162489396	0.160321476	0.158153937	0.157726954	0.160042088				
L_a, H	0.000282	0.000280223	0.000280544	0.000281633	0.000319376	0.000321892				
K_b , V/(rad/sec)	0.0503	0.048873126	0.049036542	0.049062757	0.049053501	0.049137374				
K _t , Nm/Amps	0.05292	0.051867299	0.051481376	0.050617986	0.050746409	0.050848568				
J, kg-m ²	0.00046	0.000469412	0.000456831	0.000463522	0.000460441	0.000475976				
B, Nm/(rad/sec)	0.0001778	0.000246097	0.000231629	0.000218973	0.000226532	0.000197222				

TABLE III Average of Estimated PMDC Motor Parameters for 30 Runs

In order to analyze the estimated parameters, the load test is conducted on 24 V, 3 A, 320 W, 4600 rpm PMDC motor. From load test, armature current, voltage, and speed are noted for various loading conditions. The experimental setup for load test on PMDC motor with LEM LV 25-P voltage transducer, LEM LA 55-P current transducer, MOC 7811 optoisolator, and NI USB-6221 DAQ is shown in Fig. 7. The PMDC wheelchair motor is operated with two 12 V battery. The signals are acquired at a sampling rate of 100 Hz. The instantaneous armature current and voltage variation from no load to full load and to no load are shown in Fig. 8.



Fig. 7. Experimental setup



In Matlab/Simulink angular speed and current are obtained using estimated parameter for various loading conditions from different techniques. It has been found that experimental test gives more than 5% of armature current error and more than 1% of speed error compared with bio-inspired optimization techniques. Further, standard PSO, dynamic PSO with constant inertia weight and ant colony optimization algorithm gives current error more than 1% and less than 2%. However, dynamic PSO with varying inertia weight and artificial bee colony gives speed as well as current error less than 0.5%. Figure 9 shows the percentage errors of angular speed and armature current from estimated parameters.



Fig. 9. Percentage error of angular speed and armature current

8

In order to test the significance of speed and current performance, ANOVA test was applied with significance level of 0.05. The ANOVA test shows the significant difference in speed as well as current performance between experimental and bio-inspired optimization techniques. However, there is no significant difference in speed performance among bioinspired optimization algorithms. The dynamic PSO with varying inertia weight and artificial bee colony have no significant difference with current error. However, dynamic PSO with varying inertia weight and artificial bee colony shows significant difference with current performance compared to other techniques.

The ant colony optimization algorithm needs a less number of iterations compared to other methods. However, current error is high. Dynamic PSO with constant inertia weight as well as standard PSO takes a number of iterations for convergence. However, dynamic PSO with varying inertia weight needs less iteration compared to other PSO techniques considered. Further, it is clear artificial bee colony algorithms takes less number of iteration for convergence and performance is significantly closed to dynamic PSO with varying inertia weight algorithms. Therefore, artificial bee colony algorithm may be considered for less computation and less error in the estimation of PMDC motor parameter estimation.

V. CONCLUSION

The parameter estimated from experimental tests under steady-state conditions show less accuracy with the experimental load test data. This is due to the fact that the experimental tests are unable to capture the non-linear dynamics in a motor parameter due to various influences. In this paper, applications of the standard PSO, dynamic PSO with constant inertia weight and dynamic PSO with varying inertia weight algorithms, ABC, and ACO algorithms have been studied for parameter estimation of a PMDC motor along with experimental tests. The dynamic PSO algorithm a variant of standard PSO, modifying parameter iteratively improves the parameter estimation accuracy. It is evident that the dynamic PSO with varying inertia and artificial bee colony algorithms may be used to obtain motor parameter with more accuracy without being trapped in local minima. The artificial bee colony algorithm may be preferred for estimation of PMDC motor parameters used in a wheelchair due to faster convergence as well as relatively less current error except for dynamic PSO with varying inertia. The dynamic PSO with varying inertia weight may be used for more accurate PMDC motor parameter estimation.

Appendix

Specification of wheelchair PMDC motor, 24 V, 3 A, 320 W and 4600 rpm, Motion Technology Electric & Machinery Co., Ltd., Taiwan.

ACKNOWLEDGMENT

Thanks to the Science and Engineering Research Board (SERB), Government of India, who has funded research project number: SB/FTP/ETA-54/2013. Also, thanks to the VIT University, Vellore for all other support. Authors, also thank the reviewers for their valuable suggestions to improve the paper.

REFERENCES

- D. Ding and R. A. Cooper, "Electric powered wheelchairs," *IEEE Control Systems Magazine*, vol. 25, no. 2, pp. 22–34, Apr. 2005.
- [2] S. Oh and Y. Hori, "Disturbance attenuation control for power-assist wheelchair operation on slopes", *IEEE Transactions on Control Systems Technology*, vol. 22, no. 3, pp. 828-837, 2014.
- [3] Y. P. Yang, H. C Lin, F. C. Tsai, C. T. Lu and K. H. Tu, "Design and integration of dual power wheels with rim motors for a powered wheelchair," *IET Electric Power Applications*, vol. 6, Iss. 7, pp. 419– 428, January 2012.
- [4] R. X. Chen, L. G. Chen and L. Chen, "System design consideration for digital wheelchair controller," *IEEE Transactions on Industrial Electronics*, vol. 47, no. 4, pp. 898–907, August 2000.
- [5] R. A. Cooper, L. M. Widman, D. K. Jones, R. N. Robertson, and J. F. Ster, III, "Force sensing control for electric powered wheelchairs," *IEEE Transactions on Control Systems Technology*, vol. 8, no. 1, pp. 112-117, January 2000.
- [6] K. E. Brown, R. M. Inigo and B. W. Johnson, "Design, implementation, and testing of an adaptable optimal controller for an electric wheelchair," *IEEE Transactions on Industry Applications*, vol. 26, no. 6, pp. 1144-1157, November/December 1990.
- [7] H. Seki, and N. Tanohata, "Fuzzy control for electric power-assisted wheelchair driving on disturbance roads," *IEEE Transactions on Systems, Man, and Cybernetics - Part C: Applications and Reviews*, vol. 42, no. 6, pp. 1624-1632, November 2012.
- [8] T. Shibata and T. Murakami. "Power-assist control of pushing task by repulsive compliance control in electric wheelchair", *IEEE Transactions* on *Industrial Electronics*, vol. 59, no. 1, pp. 511-520, 2012.
- [9] T. N. Nguyen, S. W. Su and H. T. Nguyen, "Robust neuro-sliding mode multivariable control strategy for powered wheelchairs," *IEEE Transactions on Neural Systems and Rehabilitation Engineering*, vol. 19, no. 1, pp. 105-111, February 2011.
- [10] K. S. Narendra and K. Parthasarathy, "Identification and control of dynamical systems using neural networks," *IEEE Transactions on Neural Networks*, vol. 1, no. 1, pp. 4-27, March 1990.
- [11] W. Lord and J. H. Hwang, "DC servomotors-modelling and parameter determination," *IEEE Transactions on Industry Applications*, vol. IA-13, no. 3, pp. 234-243, May/June 1977.
- [12] A. Rubaai and R. Kotaru, "Online identification and control of a dc motor using learning adaptation of neural networks," *IEEE Transactions* on *Industry Applications*, vol. 36, no. 3, pp. 935–942, May/June 2000.
- [13] S. Weerasooriya and M. A. El-Sharkawi, "Identification and control of a DC motor using back-propagation neural networks," *IEEE Transactions* on *Energy Conversion*, vol. 6, no. 4, pp. 663-669, December 1991.
- [14] M. Hadef and M. R. Mekideche, "Parameter identification of a separately excited dc motor via inverse problem methodology," *Turk Journal of Electrical Engineering and Computer Sciences*, vol. 17, no. 2, pp. 99-106, 2009.
- [15] D. Sendrescu, "Parameter identification of a dc motor via distribution based approach," 17th IEEE International Conference on Methods and Models in Automation and Robotics, pp. 17-22, 2012.
- [16] J. C. Basilio and M. V. Moreira, "State-space parameter identification in a second control laboratory," *IEEE Transactions on Education*, vol. 47, no. 2, pp. 204–210, May 2004.
- [17] R. Garrido and R. Miranda, "DC servomechanism parameter identification: A closed loop input error approach," *ISA Transactions*, vol. 51, pp. 42–49, 2012.
- [18] G. Mamani, J. Becedas and V. Feliu-Batlle, "On-line fast algebraic parameter and state estimation for a dc motor applied to adaptive control," *in Proceedings of the World Congress on Engineering*, vol. 2, July 2008.
- [19] S. S. Saab and R. Abi Kaed-Bey, "Parameter identification of a DC motor: an experimental approach," The 8th IEEE International

Conference on Electronics, Circuits and Systems, vol. 2, pp. 981-984, 2001.

- [20] R. A. Schulz, "A frequency response method for determining the parameters of high-performance DC motors". *IEEE Transactions on Industrial Electronics*, vol. IE-30, no. 1, pp. 39-42, February 1983.
- [21] J. Y. Hung, "Parameter estimation using sensitivity points: tutorial and experiment," *IEEE Transactions on Industrial Electronics*, vol. 48, no. 6, pp. 1043-1047, December 2001.
- [22] M. A. Obeidat, L. Y. Wang, and F. Lin, "Real-time parameter estimation of PMDC motors using quantized sensors," *IEEE Transactions on Vehicular Technology*, vol. 62, no. 7, pp. 2977-2986, September 2013.
- [23] H. Zhou, "Model-based regulator in PMDC Motor," *IEEE International conference on Systems, Man and Cybernetics*, pp. 508-512, 2004.
 [24] A. Dupuis, M. Ghribi and A. Kaddouri, "Multi-objective genetic
- [24] A. Dupuis, M. Ghribi and A. Kaddouri, "Multi-objective genetic estimation of DC motor parameters and load torque," *IEEE International Conference on Industrial Technology*, vol. 3, pp. 1511-1514, 2004.
- [25] S. Udomsuk, K-L. Areerak, K-N. Areerak and A. Srikaew, "Parameters identification of separately excited dc motor using adaptive tabu search technique," *IEEE International conference on Advances in Energy Engineering*, pp. 48-51, 2010.
- [26] D.C. Huynh and M.W. Dunnigan, "Parameter estimation of an induction machine using advanced particle swarm optimization algorithms," *IET Electric Power Applications*, vol. 4, Iss. 9, pp. 748–760, December 2010.
- [27] W. Hu and Y. Tan, "Prototype generation using multiobjective particle swarm optimization for nearest neighbor classification," *IEEE Transactions on Cybernetics*, vol. 46, no. 12, pp. 2719-2731, December 2016.
- [28] T.D. Cong and W. Zhijian, "Adaptive multi-layer particle swarm optimization with neighborhood search," *Chinese Journal of Electronics*, vol.25, no.6, pp. 1079–1088, Nov. 2016.
- [29] D. H. Tungadio, B. P. Numbi, M. W. Siti and A. A. Jimoh, "Particle swarm optimization for power system state estimation," *Neuro computing*, vol. 148, pp. 175–180, 2015.
- [30] A. Khare and S. Rangnekar, "A review of particle swarm optimization and its applications in Solar Photovoltaic system," *Applied Soft Computing*, vol. 13 pp. 2997–3006, 2013.
- [31] A. Ratanweera, S.K. Halgamuge and H.C. Watson, "Self-organizing hierarchical particle swarm optimizer with time-varying acceleration coefficients," *IEEE Transactions on Evolutionary Computation*, vol. 8, no. 3, pp. 240–255, June 2004.
- [32] W. Gu, Y. Yu and W. Hu, "Artificial bee colony algorithm-based parameter estimation of fractional-order chaotic system with time delay," *IEEE/CAA Journal of Automatica Sinica*, vol. 4, no. 1, pp. 107-113, January 2017.
- [33] M. Jamadi and F.M. Bayat, "New method for accurate parameter estimation of induction motors based on artificial bee colony algorithm," *Cornell University Library*, 2014.
- [34] F. Kang, J. Li and H. Li, "Artificial bee colony algorithm and pattern search hybridized for global optimization," *Applied Soft Computing*, vol. 13, no. 4, pp. 1781–1791, 2013.
- [35] W. Gao and S. Liu, "A modified artificial bee colony algorithm," *Computers & Operations Research*, vol. 39, pp. 687–697, 2012.
- [36] D. Karaboga and B. Basturk, "On the performance of artificial bee colony (ABC) algorithm," *Applied Soft Computing*, vol. 8, no. 1, pp. 687-697, January 2008.
- [37] V.P. Sakthivel, R. Bhuvaneswari and S. Subramanian, "Non-intrusive efficiency estimation method for energy auditing and management of inservice induction motor using bacterial foraging algorithm," *IET Electric Power Applications*, vol. 4, Iss. 8, pp. 579 - 590, 2010.
- [38] Z. Chen, Y. Zhong and J. Li, "Parameter identification of induction motors using ant colony optimization," *IEEE Congress on Evolutionary Computation*, pp. 1611 - 1616, 2008.
- [39] M. Dorigo, V. Maniezzo and A. Colorni, "Ant System: Optimization by a colony of cooperating agents," *IEEE Transactions on Systems, Man,* and Cybernetics-Part B Cybernetics, vol. 26, no. 1, pp. 29-41, February 1996.
- [40] A. Colorni, M. Dorigo and V. Maniezzo, "Distributed optimization by ant colonies," in *Proceedings of ECAL91 - European Conference on Artificial Life*, Elsevier Publishing, pp. 134-142, 1991.

NOMENCLATURE

10

torque present at the stand-still condition of the motor Α weight to mean square current error а cognitive and social parameters between 0 and 2 $c_{1.} c_{2}$ initial and final values of cognitive parameter c_{IIv}, c_{IFv} initial and final values of social parameter C2h, C2Fv number of motor parameters to be optimized d back emf in Volts e_b f viscous friction coefficient in Nm/(rad/sec) best objective function value for ant fbest global best position from a swarm in j^{th} dimension gbest k Η number of ants i_{arn} armature current in Amps particle of the swarm i Jmoment of inertia in kg-m² dimension of a particle in parameter space K_b back-emf constant in V/(rad/sec) K_t torque constant in Nm/A current iteration of the algorithm k L armature inductance in Henry М swarm size Ν number of measured samples best position of the particle i in jth dimension from own pbest " behavior probability among the H ants travels to a particular q^{th} node at p_{qs} the sth level Q quantity of the pheromone placed by the ant for each iteration number of nodes q \bar{R}_{arm} armature resistance in Ohms two independent random sequences between 0 and 1 rand₁, rand₂ S_p number of food sources number of levels T_e electromagnetic torque in Nm T_l load torque in Nm t time in seconds input vector и armature voltage in Volts Varn velocity of an i^{th} particle in j^{th} dimension parameter space at v_{ij}^k iteration k initial and final values of the inertia weight $W_{I_1} W_2$ weighting function also known as inertia parameter w position of an i^{th} particle in j^{th} dimension parameter space at x_{ij}^k iteration k output vector y estimated output vector у z parameter vector z estimated parameter vector α control the relative importance of the pheromone evaporation rate ρ au_{qs} amount of pheromone at a q^{th} node for s^{th} level angular speed in rad/sec ω d∞/dt slope of retardation curve



V. Sankardoss graduated in Electrical and Electronics Engineering from K.S. Rangasamy College of Technology, University of Madras, Tamil Nadu, India in 1999. He received M.Tech in Electrical Drives and Control from Pondicherry Engineering College, Pondicherry University, Pondicherry, India in 2005. Since 2007, he is with the

School of Electrical Engineering, VIT University, Vellore, and also pursuing the Ph. D degree from VIT University, Vellore, Tamil Nadu. His research interest is Electrical Machines, Electric wheelchair, Control Systems, Solar energy and Wind energy.



P. Geethanjali received her B.E. degree in Electrical and Electronics Engineering from Vellore Engineering College, University of Madras, Tamil Nadu, India in 2001. She obtained M. Tech in Electrical Drives and Control from Pondicherry Engineering College, Pondicherry University, Pondicherry, India in

2004. She received her Ph. D degree from VIT University, Vellore, India in 2012. Her Ph.D. thesis has been nominated for "Best Thesis" by Indian National Academy of Engineering (INAE). She has received Fulbright-Nehru Academic and Professional Excellence Fellowship during 2014-15.

Since 2004, she is with the School of electrical Engineering, VIT University, Vellore. She received grants from the Department of Science and Technology (DST), Government of India, for research on Multichannel Myoelectric Signal Studies for Actuation of Prosthetic Devices as well as Brain Controlled Wheelchair Study for Neuromuscular Impairments. Her research interests include bio-signal processing, development of assistive devices and application of renewable energy in assistive devices.